

Sub-Predictors and Classical Predictors for Finite-Dimensional Observer-Based Control of Parabolic PDEs

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Abstract—We study constant input delay compensation by using finite-dimensional observer-based controllers in the case of the 1D heat equation. We consider Neumann actuation with nonlocal measurement and employ modal decomposition with $N + 1$ modes in the observer. We introduce a chain of M sub-predictors that leads to a closed-loop ODE system coupled with infinite-dimensional tail. Given an input delay r , we present LMI stability conditions for finding M and N and the resulting exponential decay rate and prove that the LMIs are always feasible for any r . We also consider a classical observer-based predictor and show that the corresponding LMI stability conditions are feasible for any r provided N is large enough. A numerical example demonstrates that the classical predictor leads to a lower-dimensional observer. However, it is known to be hard for implementation due to the distributed input signal.

Index Terms—Distributed parameter systems, observer-based control, time-delay.

I. INTRODUCTION

FINITE-DIMENSIONAL observer-based controllers for parabolic systems were designed by the modal decomposition approach in [1], [2], [3], [4], [5]. Recently, the first constructive LMI-based method for finite-dimensional observer-based controller was suggested in [6] for the 1D heat equation under nonlocal or Dirichlet actuation and nonlocal measurement. The observer dimension N and the resulting exponential decay rate were found from simple LMI conditions. Finite-dimensional observer-based control of the Kuramoto-Sivashinsky equation with boundary actuation and point measurement was studied in [7].

Robustness with respect to small delays and/or sampling intervals for the heat equation was studied in [8], [9] for distributed static output-feedback control, in [10] for boundary state-feedback and in [11], [12] for boundary

controller based on PDE observer. Delayed implementation of finite-dimensional observer-based controllers for the 1D heat equation was introduced in [13], where in case of time-varying output delay, a combination of Lyapunov functionals with Halanay's inequality appeared to be an efficient tool.

To compensate large input/output delay, there are two main predictor methods: the classical predictor, which is based on a reduction approach [14] or the backstepping approach [15] and sub-predictors or chain of observers [16], [17], [18], [19]. The classical predictors for state-feedback control of PDEs were suggested in [15], [20], [21]. For the heat equation, a PDE sub-predictor (an observer of the future state) was presented in [11]. A chain of observers for the estimation of heat equation with a large output delay was designed in [22].

In the recent paper [23], reduced-order LMI stability conditions were introduced for finite-dimensional observer-based control. This was presented for the heat equation with Neumann actuation and non-local measurement. The dimension of the LMIs does not grow with the dimension of the observer N . Moreover, feasibility of the LMIs for N implies their feasibility for $N + 1$. In [23], the classical predictor was extended to finite-dimensional observer-based control. This predictor compensated delay in the finite-dimensional controller, whereas the infinite-dimensional part still depended on the large input delay. It was shown in a numerical example that the predictor allows for larger delays. However, the feasibility of LMIs for arbitrary delays was not proved due to complexity of the analysis in the presence of time-varying output delay.

The present paper is dedicated to predictor methods for finite-dimensional observer-based control of parabolic PDEs with constant input delay r . As in [23], we consider the 1D heat equation under Neumann actuation and non-local measurement. The main novelty is in use of sub-predictors for such a system. We show that for any r there exists a chain of M sub-predictors and a large enough number of modes $N + 1$ employed in observer that guarantee the stability of the closed-loop system. We present LMI stability conditions for finding M , N and the resulting exponential decay rate. We prove that these LMIs are always feasible for all r and large enough M and N . We also consider the classical predictor which compensates the delay in the finite-dimensional part, as introduced in [23] (if the time-varying input/output delays are omitted). This is the first time that feasibility guarantees for the resulting LMIs with arbitrary delays are proved

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for both sub-predictors and predictors. This proof is challenging, due to coupling in the closed-loop system. A numerical example demonstrates that for the same N , the classical predictor allows larger delays found from the LMIs, whereas for the same delay they employ lower-dimensional observers than the sub-predictors. However, as is well-known [24], [25], they are harder to implement, due to the distributed input term which should be carefully discretized. This letter is an essential step towards the use of sub-predictors and classical predictors for delay compensation in PDEs, via finite-dimensional observers.

Notations and preliminaries: $L^2(0, 1)$ is the Hilbert space of Lebesgue measurable and square integrable functions $f : [0, 1] \rightarrow \mathbb{R}$ with the inner product $\langle f, g \rangle := \int_0^1 f(x)g(x)dx$ and induced norm $\|f\|^2 := \langle f, f \rangle$. $H^1(0, 1)$ is the space of functions $f : [0, 1] \rightarrow \mathbb{R}$ with square integrable weak derivative, with the norm $\|f\|_{H^1}^2 := \sum_{j=0}^1 \|f^{(j)}\|^2$. The Euclidean norm on \mathbb{R}^n is denoted by $|\cdot|$. For $P \in \mathbb{R}^{n \times n}$, $P > 0$ means that P is symmetric and positive definite. The sub-diagonal elements of a symmetric matrix will be denoted by $*$. \otimes is the standard Kronecker product. For $U \in \mathbb{R}^{n \times n}$, $U > 0$ and $x \in \mathbb{R}^n$ let $|x|_U^2 = x^T U x$. \mathbb{Z}_+ is the set of nonnegative integers.

Recall that the Sturm-Liouville eigenvalue problem

$$\phi'' + \lambda\phi = 0, \quad x \in [0, 1], \quad \phi'(0) = \phi'(1) = 0, \quad (1)$$

induces a sequence of eigenvalues $\lambda_n = n^2\pi^2$, $n \in \mathbb{Z}_+$ with corresponding eigenfunctions

$$\phi_0(x) = 1, \quad \phi_n(x) = \sqrt{2} \cos(\sqrt{\lambda_n}x), \quad n \geq 1. \quad (2)$$

The eigenfunctions form a complete orthonormal system in $L^2(0, 1)$. Given $N \in \mathbb{Z}_+$ and $h \in L^2(0, 1)$ satisfying $h \stackrel{L^2}{=} \sum_{n=0}^{\infty} h_n \phi_n$ we denote $\|h\|_N^2 = \sum_{n=N+1}^{\infty} h_n^2$.

II. SUB-PREDICTORS VS CLASSICAL PREDICTORS

We consider the PDE

$$\begin{aligned} z_t(x, t) &= z_{xx}(x, t) + qz(x, t), \quad x \in [0, 1], \quad t \geq 0, \\ z_x(0, t) &= 0, \quad z_x(1, t) = u(t-r) \end{aligned} \quad (3)$$

under delayed Neumann actuation with known delay r and non-local measurement

$$y(t) = \langle c, z(\cdot, t) \rangle, \quad t \geq 0 \quad (4)$$

with $c \in L^2(0, 1)$. To compensate the delay, we will present in this section both sub-predictors and classical predictors.

Using modal decomposition, we present the solution to (3) as

$$z(x, t) \stackrel{L^2}{=} \sum_{n=0}^{\infty} z_n(t) \phi_n(x), \quad z_n(t) = \langle z(\cdot, t), \phi_n \rangle \quad (5)$$

with ϕ_n , $n \in \mathbb{Z}_+$ given in (2). Differentiating under the integral, integrating by parts and using (1) and (2) we obtain (similar to [10] and the references therein)

$$\begin{aligned} \dot{z}_n(t) &= (-\lambda_n + q)z_n(t) + b_n u(t-r), \quad t \geq 0 \\ b_0 &= 1, \quad b_n = (-1)^n \sqrt{2}, \quad n \geq 1. \end{aligned} \quad (6)$$

Let $\delta > 0$ be a desired decay rate. Since $\lim_{n \rightarrow \infty} \lambda_n = \infty$, there exists some $N_0 \in \mathbb{Z}_+$ such that

$$-\lambda_n + q < -\delta, \quad n > N_0. \quad (7)$$

Let

$$\begin{aligned} A_0 &= \text{diag}\{-\lambda_0 + q, \dots, -\lambda_{N_0} + q\}, \\ L_0 &= [l_0, \dots, l_{N_0}]^T, \quad B_0 = [b_0, \dots, b_{N_0}]^T \\ C_0 &= [c_0, \dots, c_{N_0}], \quad c_n = \langle c, \phi_n \rangle, \quad n \in \mathbb{Z}_+. \end{aligned} \quad (8)$$

Assume that

$$c_n \neq 0, \quad 0 \leq n \leq N_0. \quad (9)$$

Then (A_0, C_0) is observable, by the Hautus lemma. We choose $L_0 = [l_0, \dots, l_{N_0}]^T$ which satisfies the following Lyapunov inequality:

$$P_0(A_0 - L_0 C_0) + (A_0 - L_0 C_0)^T P_0 < -2\delta P_0, \quad (10)$$

where $0 < P_0 \in \mathbb{R}^{(N_0+1) \times (N_0+1)}$.

Similarly, by the Hautus lemma, $b_n \neq 0, n \in \mathbb{Z}_+$ implies that (A_0, B_0) is controllable. Let $K_0 \in \mathbb{R}^{1 \times (N_0+1)}$ satisfy

$$P_c(A_0 - B_0 K_0) + (A_0 - B_0 K_0)^T P_c < -2\delta P_c, \quad (11)$$

where $0 < P_c \in \mathbb{R}^{(N_0+1) \times (N_0+1)}$.

In our finite-dimensional observer-based predictor design, the closed-loop system will be presented as a coupled system of ODEs and the infinite-dimensional tail. This complicates the proof of stabilization for all $r > 0$ under higher-dimensional observers.

Given $N \geq N_0$ denote

$$\begin{aligned} \hat{z}^{N_0}(t) &= \text{col}\{\hat{z}_i\}_{i=0}^{N_0}, \quad \hat{z}^{N-N_0}(t) = \text{col}\{\hat{z}_i\}_{i=N_0+1}^N, \\ A_1 &= \text{diag}\{-\lambda_{N_0+1} + q, \dots, -\lambda_N + q\}, \\ B_1 &= [b_{N_0+1}, \dots, b_N]^T, \quad C_1 = [c_{N_0+1}, \dots, c_N]. \end{aligned} \quad (12)$$

A. Sub-Predictors

In order to deal with a large delay r , we subdivide r into M parts of equal size $\frac{r}{M}$, where $M \in \mathbb{Z}_+$, $M \geq 1$. We first consider $M \geq 2$ and employ a chain of sub-predictors (observers of the future state)

$$\hat{z}_1^{N_0}(t-r) \mapsto \dots \mapsto \hat{z}_i^{N_0}\left(t - \frac{M-i+1}{M}r\right) \mapsto \dots \mapsto \hat{z}^{N_0}(t). \quad (13)$$

Here $\hat{z}_i^{N_0}(t - \frac{M-i+1}{M}r) \mapsto \hat{z}_{i+1}^{N_0}(t - \frac{M-i}{M}r)$ means that $\hat{z}_i^{N_0}(t)$ predicts the value of $\hat{z}_{i+1}^{N_0}(t + \frac{r}{M})$. Similarly, $\hat{z}_M^{N_0}(t)$ predicts the value of $\hat{z}^{N_0}(t + \frac{r}{M})$. The sub-predictors satisfy the following ODEs for $t \geq 0$

$$\begin{aligned} \dot{\hat{z}}_M^{N_0}(t) &= A_0 \hat{z}_M^{N_0}(t) + B_0 u\left(t - \frac{M-1}{M}r\right) \\ &\quad - L_0 \left[C_0 \hat{z}_M^{N_0}\left(t - \frac{r}{M}\right) + C_1 \hat{z}^{N-N_0}(t) - y(t) \right], \\ \dot{\hat{z}}_i^{N_0}(t) &= A_0 \hat{z}_i^{N_0}(t) + B_0 u\left(t - \frac{i-1}{M}r\right) \\ &\quad - L_0 C_0 \left[\hat{z}_i^{N_0}\left(t - \frac{r}{M}\right) - \hat{z}_{i+1}^{N_0}(t) \right], \quad 1 \leq i \leq M-1, \\ \hat{z}_i^{N_0}(t) &= 0, \quad t \leq 0, \quad 1 \leq i \leq M, \end{aligned} \quad (14)$$

whereas $\hat{z}^{N-N_0}(t)$ satisfies the following ODE

$$\dot{\hat{z}}^{N-N_0}(t) = A_1 \hat{z}^{N-N_0}(t) + B_1 u(t-r), \quad \hat{z}^{N-N_0}(t) = 0, \quad t \leq 0. \quad (15)$$

The finite-dimensional observer $\hat{z}(x, t)$ of the state $z(x, t)$, based on $(M-1)(N_0+1) + N + 1$ -dimensional system of ODEs (14)-(15), is given by

$$\begin{aligned} \hat{z}(x, t) &= \hat{z}_1^{N_0}(t-r) \cdot \text{col}\{\phi_j(x)\}_{j=0}^{N_0} \\ &\quad + \hat{z}^{N-N_0}(t) \cdot \text{col}\{\phi_j(x)\}_{j=N_0+1}^N. \end{aligned} \quad (16)$$

The controller is further chosen as

$$u(t) = -K_0 \hat{z}_1^{N_0}(t). \quad (17)$$

In particular, (14) implies $u(t) = 0$ for $t \leq 0$.

For well-posedness we introduce the change of variables $w(x, t) = z(x, t) - \frac{1}{2}x^2u(t-r)$. Then, the closed-loop system is presented as

$$\begin{aligned} w_t(x, t) &= w_{xx}(x, t) + qw(x, t) + f(x, t), \quad x \in [0, 1], \quad t \geq 0, \\ w_x(0, t) &= 0, \quad w_x(1, t) = 0, \\ f(x, t) &= -\frac{1}{2}x^2\dot{u}(t-r) + \left(\frac{q}{2}x^2 + 1\right)u(t-r), \end{aligned} \quad (18)$$

the ODEs (14) and (17). Let $z(\cdot, 0) = w(\cdot, 0) \in H^1(0, 1)$. We apply the step method on $\{[jr, (j+1)r]\}_{j=0}^\infty$. For $t \in [0, r]$ we have that $f(x, t) \equiv 0$. By [26, Ths. 6.3.1 and 6.3.3], (18) has a unique classical solution $z = w \in C([0, r], L^2(0, 1)) \cap C^1((0, r], L^2(0, 1))$ such that $w(\cdot, t) \in H^2(0, 1)$ with $w_x(0, t) = w_x(1, t) = 0$ for $t \in (0, r]$. Furthermore, since $u(t-r) \equiv 0$ for $t \in [0, r]$, (15) implies that $\hat{z}^{N-N_0}(t) \in C^1[0, r]$. Since $y \in C[0, r]$, considering (14) on the subintervals $\{[\frac{j}{M}r, \frac{(j+1)}{M}r]\}_{j=0}^{M-1}$, it can be seen that $\hat{z}_i^{N_0} \in C^1[0, r]$, $1 \leq i \leq M$. Furthermore, $\hat{z}_1^{N_0}$ is Lipschitz for $t \in [0, r]$. Next, we consider $t \in [r, 2r]$. Since $\hat{z}_1^{N_0}(t) \in C^1[0, r]$, with $\hat{z}_1^{N_0}(t)$ Lipschitz on $[0, r]$, we have that $f(x, t)$ is Lipschitz on $[r, 2r]$. By [26, Ths. 6.3.1 and 6.3.3], (18) has a unique classical solution for $t \in [r, 2r]$. Continuing step-by-step and using $z(x, t) = w(x, t) + \frac{1}{2}x^2u(t-r)$, (3) has a unique solution $z \in C([0, \infty), L^2(0, 1)) \cap C^1((0, \infty) \setminus \mathcal{J}, L^2(0, 1))$, where $\mathcal{J} = \{\frac{j}{M}r\}_{j=0}^\infty$. Moreover, $z(\cdot, t) \in H^2(0, 1)$ with $z_x(0, t) = 0$, $z_x(1, t) = u(t-r)$ for $t \in [0, \infty)$.

Define the estimation errors for $1 \leq i \leq M-1$ as follows:

$$\begin{aligned} e^{N-N_0}(t) &= \text{col}\{z_{N_0+1}(t), \dots, z_N(t)\} - \hat{z}^{N-N_0}(t), \\ e_M^{N_0}(t) &= z^{N_0}(t) - \hat{z}_M^{N_0}\left(t - \frac{1}{M}r\right), \\ e_i^{N_0}(t) &= \hat{z}_{i+1}^{N_0}\left(t - \frac{M-i}{M}r\right) - \hat{z}_i^{N_0}\left(t - \frac{M-i+1}{M}r\right). \end{aligned} \quad (19)$$

From (14) and (19) we have

$$\hat{z}_1^{N_0}(t-r) + \sum_{i=1}^M e_i^{N_0}(t) = z^{N_0}(t). \quad (20)$$

In particular, if the errors $e_i^{N_0}(t)$, $1 \leq i \leq M$ converge to zero, we have $\hat{z}_1^{N_0}(t) \rightarrow z^{N_0}(t+r)$, meaning that $\hat{z}_1^{N_0}(t)$ sequentially forecasts the future system state $z^{N_0}(t+r)$. Using (4), (8) and (19), the innovation term in the ODE for $\hat{z}_M^{N_0}(t)$ (see (14)), can be presented as

$$\begin{aligned} C_0\hat{z}_M^{N_0}\left(t - \frac{r}{M}\right) + C_1\hat{z}^{N-N_0}(t) - y(t) &= -C_0e_M^{N_0}(t) \\ -C_1e^{N-N_0}(t) - \zeta(t), \quad \zeta(t) &= \sum_{n=N+1}^\infty c_n z_n(t). \end{aligned} \quad (21)$$

By the Young inequality we have

$$\zeta^2(t) \leq \|c\|_N^2 \sum_{n=N+1}^\infty z_n^2(t). \quad (22)$$

Using (6), (14) and (21) we obtain the following dynamics of the estimation errors for $t \geq 0$

$$\begin{aligned} \dot{e}_M^{N_0}(t) &= A_0e_M^{N_0}(t) - L_0C_0e_M^{N_0}\left(t - \frac{r}{M}\right) \\ &\quad - L_0C_1e^{N-N_0}\left(t - \frac{r}{M}\right) - L_0\zeta\left(t - \frac{r}{M}\right), \\ \dot{e}_{M-1}^{N_0}(t) &= A_0e_{M-1}^{N_0}(t) - L_0C_0e_{M-1}^{N_0}\left(t - \frac{r}{M}\right) \end{aligned}$$

$$\begin{aligned} &+ L_0C_0e_M^{N_0}\left(t - \frac{r}{M}\right) + L_0C_1e^{N-N_0}\left(t - \frac{r}{M}\right) \\ &+ L_0\zeta\left(t - \frac{r}{M}\right), \end{aligned}$$

$$\begin{aligned} \dot{e}_i^{N_0}(t) &= A_0e_i^{N_0}(t) - L_0C_0e_i^{N_0}\left(t - \frac{r}{M}\right) \\ &+ L_0C_0e_{i+1}^{N_0}\left(t - \frac{r}{M}\right), \quad 1 \leq i \leq M-2 \end{aligned} \quad (23)$$

and

$$\dot{e}^{N-N_0}(t) = A_1e^{N-N_0}(t). \quad (24)$$

From (6), (17) and (20), $z^{N_0}(t)$ satisfies

$$\dot{z}^{N_0}(t) = (A_0 - B_0K_0)z^{N_0}(t) + B_0K_0 \sum_{i=1}^M e_i^{N_0}(t). \quad (25)$$

We introduce the notations

$$\begin{aligned} X_e(t) &= \text{col}\{e_i^{N_0}(t)\}_{i=1}^M, \quad v_e(t) = X_e\left(t - \frac{r}{M}\right) - X_e(t), \\ F_e &= \text{diag}\{I_{M-1} \otimes (A_0 - L_0C_0) + J_{M-1}(0) \otimes L_0C_0, A_0 - L_0C_0\}, \\ G_e &= \text{diag}\{I_{M-1} \otimes (-L_0C_0) + J_{M-1}(0) \otimes L_0C_0, -L_0C_0\}, \\ \mathcal{L}_e &= \text{col}\{0_{(M-2)(N_0+1) \times 1}, L_0, -L_0\}, \\ \mathcal{K}_e &= [K_0, \dots, K_0] \in \mathbb{R}^{1 \times M(N_0+1)} \end{aligned}$$

where $J_{M-1}(0)$ is a Jordan block of order $M-1$ with zero diagonal. Note that (24) implies $e^{N-N_0}\left(t - \frac{r}{M}\right) = e^{-A_1\frac{r}{M}}e^{N-N_0}(t)$. Then, using (6), (17), (23) and (25), the reduced-order (i.e., decoupled from $\hat{z}^{N-N_0}(t)$) closed-loop system can be presented as

$$\begin{aligned} \dot{z}^{N_0}(t) &= (A_0 - B_0K_0)z^{N_0}(t) + B_0\mathcal{K}_eX_e(t) \\ \dot{X}_e(t) &= F_eX_e(t) + G_e v_e(t) + \mathcal{L}_e\zeta\left(t - \frac{r}{M}\right) \\ &\quad + \mathcal{L}_eC_1e^{-A_1\frac{r}{M}}e^{N-N_0}(t), \\ \dot{z}_n(t) &= (-\lambda_n + q)z_n(t) - b_nK_0z^{N_0}(t), \\ &\quad + b_n\mathcal{K}_eX_e(t), \quad n > N. \end{aligned} \quad (26)$$

In the case $M=1$, $\hat{z}^{N_0}(t)$ satisfies the first ODE in (14) and predicts $z^{N_0}(t+r)$. Here $X_e(t) = e_1^{N_0}(t)$ and the closed-loop system has the form (24) and (26), where now

$$F_e = A_0 - L_0C_0, \quad G_e = -L_0C_0, \quad \mathcal{K}_e = K_0, \quad \mathcal{L}_e = -L_0.$$

Differently from the existing finite-dimensional controllers [6], [13], where the closed-loop systems is written in terms of the observer and the tail $z_n(t)$ ($n > N$), here (26) is presented in terms of the state $z^{N_0}(t)$, the estimation errors $X_e(t)$ and the tail. This allows to eliminate the delay r from the ODEs of $z^{N_0}(t)$ and $z_n(t)$, $n > N$ while decreasing it to $\frac{r}{M}$ (which is small for large M) in the ODE of $X_e(t)$.

Remark 1: In the case of sub-predictors for linear ODEs, the closed-loop system is given by (23) and (25), where $\zeta = 0$ and $e^{N-N_0} = 0$. Thus, exponential stability of

$$\dot{e}_M^{N_0}(t) = (A_0 - L_0C_0)e_M^{N_0}(t) - L_0C_0v_{e,M}(t), \quad (27)$$

where $v_{e,M}(t) = e_M^{N_0}\left(t - \frac{r}{M}\right) - e_M^{N_0}(t)$, guarantees the stability of the closed-loop system due to ISS of the $e_i^{N_0}$ ($1 \leq i \leq M-1$) systems with respect to $e_{i+1}^{N_0}$. This is different from the infinite-dimensional closed-loop system (26), where the finite-dimensional part of the system is coupled via $\zeta(t)$ with the infinite-dimensional tail z_n ($n > N$). Here the proof of stabilization for any delay $r > 0$ provided M and N are large enough becomes challenging.

For L^2 -stability analysis of (24) and (26) we define the Lyapunov functional

$$\begin{aligned}
V(t) &:= V_0(t) + V_e(t) + V_Q(t) + p_e |e^{N-N_0}(t)|^2, \\
V_0(t) &= |z^{N_0}(t)|_{P_0}^2 + \sum_{n=N+1}^{\infty} z_n^2(t), \\
V_Q(t) &= Q \int_{t-\frac{r}{M}}^t e^{-2\delta(t-s)} \zeta^2(s) ds, \\
V_e(t) &= |X_e(t)|_{P_e}^2 + V_{S_e, X_e}(t) + V_{R_e, X_e}(t) \quad (28)
\end{aligned}$$

Here $0 < P_0 \in \mathbb{R}^{(N_0+1) \times (N_0+1)}$, $0 < Q, p_e \in \mathbb{R}$, $0 < P_e \in \mathbb{R}^{M(N_0+1) \times M(N_0+1)}$ and

$$\begin{aligned}
V_{S_e, X_e}(t) &:= \int_{t-\frac{r}{M}}^t e^{-2\delta(t-s)} |X_e(s)|_{S_e}^2 ds, \\
V_{R_e, X_e}(t) &:= \frac{r}{M} \int_{-\frac{r}{M}}^0 \int_{t+\theta}^t e^{-2\delta(t-s)} |\dot{X}_e(s)|_{R_e}^2 ds d\theta, \quad (29)
\end{aligned}$$

where $0 < S_0, R_0 \in \mathbb{R}^{M(N_0+1) \times M(N_0+1)}$. $V_0(t)$ compensates $\zeta(t)$ using (22), whereas $V_e(t)$ compensates $\frac{r}{M}$ in the ODEs of the estimation errors. Differentiation of $V_Q(t)$ gives

$$\dot{V}_Q + 2\delta V_Q = Q\zeta^2(t) - \epsilon_M Q\zeta^2\left(t - \frac{r}{M}\right), \quad \epsilon_M = e^{-2\delta \frac{r}{M}}. \quad (30)$$

Differentiating $V_0(t)$ along (26) we obtain

$$\begin{aligned}
\dot{V}_0 + 2\delta V_0 &= (z^{N_0})^T(t) [2\delta P_0 + P_0(A_0 - B_0 K_0) \\
&\quad + (A_0 - B_0 K_0)^T P_0] z^{N_0}(t) + 2(z^{N_0})^T(t) P_0 B_0 \mathcal{K}_e X_e(t) \\
&\quad + 2 \sum_{n=N+1}^{\infty} (-\lambda_n + q + \delta) z_n^2(t) \\
&\quad + 2 \sum_{n=N+1}^{\infty} z_n(t) b_n [\mathcal{K}_e X_e(t) - K_0 z^{N_0}(t)]. \quad (31)
\end{aligned}$$

Let $\alpha_1, \alpha > 0$. By the Young inequality we have

$$\begin{aligned}
2 \sum_{n=N+1}^{\infty} z_n(t) b_n [\mathcal{K}_e X_e(t) - K_0 z^{N_0}(t)] \\
\leq \left(\frac{1}{\alpha} + \frac{1}{\alpha_1}\right) \sum_{n=N+1}^{\infty} \lambda_n z_n^2(t) + \frac{2\alpha}{N\pi^2} |K_0 z^{N_0}(t)|^2 \\
+ \frac{2\alpha_1}{N\pi^2} |\mathcal{K}_e X_e(t)|^2 \quad (32)
\end{aligned}$$

where we used the value of b_n , given in (6), and the estimate

$$\sum_{n=N+1}^{\infty} \frac{b_n^2}{\lambda_n} \leq \frac{2\alpha}{\pi^2} \int_N^{\infty} \frac{dx}{x^2} = \frac{2\alpha}{N\pi^2}. \quad (33)$$

Differentiation of $V_e(t)$ and Jensen's inequality lead to

$$\begin{aligned}
\dot{V}_e + 2\delta V_e &\leq X_e^T(t) [P_e F_e + F_e^T P_e + 2\delta P_e] X_e(t) \\
&\quad + 2X_e^T(t) P_e G_e v_e(t) + 2X_e^T(t) P_e \mathcal{L}_e \zeta\left(t - \frac{r}{M}\right) \\
&\quad + 2X_e^T(t) P_e \mathcal{L}_e C_1 e^{-A_1 \frac{r}{M}} e^{N-N_0(t)} + |X_e(t)|_{S_e}^2 - \epsilon_M \\
&\quad \times \left[|X_e(t) + v_e(t)|_{S_e}^2 + |v_e(t)|_{R_e}^2 \right] + \left(\frac{r}{M}\right)^2 |\dot{X}_e(t)|_{R_e}^2. \quad (34)
\end{aligned}$$

To compensate $\zeta^2(t)$ we use monotonicity of λ_n , $n \in \mathbb{Z}_+$, (31) and (32) to obtain

$$\begin{aligned}
2 \sum_{n=N+1}^{\infty} \left(-\lambda_n + q + \delta + \left[\frac{1}{2\alpha} + \frac{1}{2\alpha_1} \right] \lambda_n \right) z_n^2(t) \\
\stackrel{(22)}{\leq} 2 \left(-\lambda_{N+1} + q + \delta + \left[\frac{1}{2\alpha} + \frac{1}{2\alpha_1} \right] \lambda_{N+1} \right) \|c\|_N^2 \zeta^2(t) \quad (35)
\end{aligned}$$

provided $-\lambda_{N+1} + q + \delta + \left[\frac{1}{2\alpha} + \frac{1}{2\alpha_1} \right] \lambda_{N+1} \leq 0$. Let

$$\eta(t) = \text{col}\{z^{N_0}(t), X_e(t), v_e(t), \zeta\left(t - \frac{r}{M}\right), e^{N-N_0}(t)\}.$$

From (31)-(35) we have

$$\dot{V}(t) + 2\delta V(t) \leq \eta^T(t) \Psi_1 \eta(t) + \frac{2}{\|c\|_N^2} \Psi_2 \zeta^2(t) \leq 0, \quad t \geq 0,$$

provided

$$\Psi_2 = -\lambda_{N+1} + q + \delta + \left[\frac{1}{2\alpha} + \frac{1}{2\alpha_1} \right] \lambda_{N+1} + \frac{Q\|c\|_N^2}{2} < 0,$$

$$\Psi_1 = \Psi_{\text{full}} + \left(\frac{r}{M}\right)^2 \Lambda^T R_e \Lambda < 0,$$

$$\Psi_{\text{full}} = \begin{bmatrix} \Psi_0 & \Sigma_1 \\ * & \Gamma_1 \end{bmatrix}, \quad \Sigma_1 = \begin{bmatrix} 0 \\ 0 \\ P_e \mathcal{L}_e C_1 e^{-A_1 \frac{r}{M}} \\ 0 \end{bmatrix},$$

$$\Gamma_1 = 2p_e [A_1 + \delta I], \quad \Lambda = [\Lambda_0, \mathcal{L}_e C_1 e^{-A_1 \frac{r}{M}}]$$

with

$$\begin{aligned}
\Psi_0 &= \begin{bmatrix} \psi & P_0 B_0 \mathcal{K}_e & 0 \\ * & \Phi(P_e, S_e, R_e) & P_e \text{col}\{\mathcal{L}_e, 0\} \\ * & * & -\epsilon_M Q \end{bmatrix} \\
&\quad + \frac{2}{N\pi^2} \text{diag}\{\alpha K_0^T K_0, \alpha_1 [\mathcal{K}_e \ 0]^T [\mathcal{K}_e \ 0], 0\}, \\
\psi &= P_0(A_0 - B_0 K_0) + (A_0 - B_0 K_0)^T P_0 + 2\delta P_0, \\
\Phi(P_e, S_e, R_e) &= \begin{bmatrix} P_e F_e + F_e^T P_e + 2\delta P_e + (1 - \epsilon_M) S & P_e G_e - \epsilon_M S_e \\ * & -\epsilon_M (S_e + R_e) \end{bmatrix}, \\
\Lambda_0 &= [0, F_e, G_e, \mathcal{L}_e]. \quad (36)
\end{aligned}$$

By Schur's complement we have that $\Psi_2 < 0$ iff

$$\begin{bmatrix} -\lambda_{N+1} + q + \delta + \frac{Q\|c\|_N^2}{2} & 1 & 1 \\ * & -2\lambda_{N+1}^{-1} & \text{diag}\{\alpha, \alpha_1\} \end{bmatrix} < 0. \quad (37)$$

Finally, that (7) yields $\Gamma_1 < 0$. Therefore, applying Schur complement and taking $p_e \rightarrow \infty$ we find that $\Psi_1 < 0$ if

$$\Psi_0 + \left(\frac{r}{M}\right)^2 \Lambda_0^T R_e \Lambda_0 < 0 \quad (38)$$

with Λ_0 given in (36). Note that (37) and (38) are reduced-order LMIs whose dimension is independent of N . Summarizing, we arrive at:

Theorem 1: Consider (3), measurement (4) with $c \in L^2(0, 1)$ satisfying (9) and control law (17). Let $\delta > 0$ be a desired decay rate. Let $N_0 \in \mathbb{Z}_+$ satisfy (7) and $N \geq N_0 + 1$. Assume that L_0 and K_0 are obtained using (10) and (11), respectively. Given $M \in \mathbb{Z}_+$, $M \geq 1$ and $r > 0$, let there exist positive definite matrices P_0, P_e, S_e, R_e and scalars $Q, \alpha, \alpha_1 > 0$ such that (37) and (38) hold. Then the solution $z(x, t)$ to (3) under the control law (17) and the corresponding subpredictor-based observer $\hat{z}(x, t)$ defined by (14), (15) and (16) satisfy

$$\|z(\cdot, t) - \hat{z}(\cdot, t)\| + \|z(\cdot, t)\| \leq D e^{-\delta t} \|z(\cdot, 0)\| \quad (39)$$

for some constant $D > 0$.

We show next that (37) and (38) are feasible for any delay $r > 0$ provided M and N are large enough. For this purpose consider (27) and

$$V_M(t) = \left| e_M^{N_0}(t) \right|_P^2 + V_{S_e, \epsilon_M}^{N_0}(t) + V_{R_e, \epsilon_M}^{N_0}(t)$$

with $V_{S_e, \epsilon_M}^{N_0}(t), V_{R_e, \epsilon_M}^{N_0}(t)$ as in (29), where $0 < P, S, R \in \mathbb{R}^{N_0+1}$. The LMI

$$\begin{aligned} \mathcal{J}(P, S, R) &= \begin{bmatrix} \phi - PL_0C_0 - e^{-2\delta h}S \\ * & -e^{-2\delta h}(S + R) \end{bmatrix} \\ &+ h^2 \begin{bmatrix} (A_0 - L_0C_0)^T \\ -C_0^T L_0^T \end{bmatrix} R [A_0 - L_0C_0 - L_0C_0] < 0, \\ \phi &= P(A_0 - L_0C_0) + (A_0 - L_0C_0)^T P + 2\delta P + (1 - e^{-2\delta h})S. \end{aligned} \quad (40)$$

with $h = \frac{r}{M}$, guarantees $\dot{V}_M(t) + 2\delta V_M(t) \leq 0$ along (27). Given $\delta > 0$, (10) implies that (40) is feasible for small enough $h > 0$ (see, e.g., [27]).

Proposition 1: Given $h > 0$, let $0 < P, S, R \in \mathbb{R}^{N_0+1}$ such that (40) holds. Then, given $r > 0$ and $M > \frac{r}{h}$, there exists some N_* such that for all $N > N_*$, (37) and (38) are feasible.

Proof: We first show that there exist $0 < P_e, S_e, R_e \in \mathbb{R}^{M(N_0+1) \times M(N_0+1)}$ such that

$$\Phi(P_e, S_e, R_e) + \left(\frac{r}{M}\right)^2 \begin{bmatrix} F_e^T \\ G_e^T \end{bmatrix} R_e [F_e \ G_e] < 0 \quad (41)$$

with $\Phi(P_e, S_e, R_e)$ given in (36). Consider the ODE

$$\dot{X}_e = F_e X_e(t) + G_e v_e(t) \quad (42)$$

obtained from (26) by setting $\zeta(t) \equiv 0$ and $e^{N-N_0}(t) \equiv 0$. For $V_e(t)$, given in (28), by standard arguments it can be easily verified that (41) guarantees $\dot{V}_e(t) + 2\delta V_e(t) \leq 0$. We will construct $V_e(t)$ recursively, by using P, S and R , thereby obtaining P_e, S_e and R_e . First, consider the ODE (27). Since $\frac{r}{M} < h$, (40) holds with h replaced by $\frac{r}{M}$. Next, consider (27) and the ODE of $e_{M-1}^{N_0}(t)$ in (42):

$$\dot{e}_{M-1}^{N_0}(t) = (A_0 - L_0C_0)e_{M-1}^{N_0}(t) - L_0C_0(v_{e,M-1}(t) - e_{M-1}^{N_0}(t)) \quad (43)$$

Let $\mu > 0$ and define

$$V_{M-1}(t) = \left| e_{M-1}^{N_0}(t) \right|_P^2 + V_{S, e_{M-1}^{N_0}}(t) + V_{R, e_{M-1}^{N_0}}(t) + \mu V_M(t)$$

where $V_M(t)$ is rescaled by μ . Using (27) and (43), the following LMI guarantees $\dot{V}_{M-1}(t) + 2\delta V_{M-1}(t) \leq 0$:

$$\begin{bmatrix} \mathcal{J}(P, S, R) & \begin{bmatrix} PL_0C_0 + \left(\frac{r}{M}\right)^2 (A_0 - L_0C_0)^T RL_0C_0 & 0 \\ -\left(\frac{r}{M}\right)^2 C_0^T L_0^T RL_0C_0 & 0 \end{bmatrix} \\ * & \mu \mathcal{J}(P, S, R) + \left(\frac{r}{M}\right)^2 \begin{bmatrix} C_0^T L_0^T RL_0C_0 & 0 \\ 0 & 0 \end{bmatrix} \end{bmatrix} < 0. \quad (44)$$

Since $\mathcal{J}(P, S, R) < 0$, taking μ large enough and applying Schur complement it can be seen that (44) holds. Repeating these arguments by backward induction, i.e., choosing

$$\begin{aligned} V_{M-j}(t) &= \left| e_{M-j}^{N_0}(t) \right|_P^2 + V_{S, e_{M-j}^{N_0}}(t) + V_{R, e_{M-j}^{N_0}}(t) \\ &+ \mu V_{M-j+1}(t), \quad 2 \leq j \leq M-1 \end{aligned} \quad (45)$$

and increasing μ at each step, we obtain that (41) holds with $W_e = \text{diag}\{\mu^j W\}_{j=0}^{M-1}$, $W \in \{P, S, R\}$. Next, recall (37) and (38), with Ψ_0 given in (36). Set $\alpha = \alpha_1 = 2$ and let $\beta > 0$. Rescaling, we replace $\Phi(P_e, S_e, R_e)$ with $\Phi(\beta P_e, \beta S_e, \beta R_e) = \beta \Phi(P_e, S_e, R_e)$. Let $P_0 = P_c$, given in (11), resulting in $\psi < 0$ in (36). Setting $\beta > 0$ to be large enough, then choosing $Q = N$ large enough and applying Schur complement twice in (38), we find that (37) and (38) hold. ■

B. Classical Observer-Based Predictor

For the case of a classical predictor, we consider a $N+1$ dimensional observer of the form

$$\begin{aligned} \hat{z}(x, t) &= \hat{z}^{N_0}(t) \cdot \text{col}\{\phi_j(x)\}_{j=0}^{N_0} \\ &+ \hat{z}^{N-N_0}(t) \cdot \text{col}\{\phi_j(x)\}_{j=N_0+1}^N. \end{aligned} \quad (46)$$

Here $\hat{z}^{N-N_0}(t)$ is defined in (12) and satisfies (15), whereas $\hat{z}^{N_0}(t)$ satisfies the following ODE

$$\begin{aligned} \dot{\hat{z}}^{N_0}(t) &= A_0 \hat{z}^{N_0}(t) + B_0 u(t-r) \\ &- L_0 [C_0 \hat{z}^{N_0}(t) + C_1 \hat{z}^{N-N_0}(t) - y(t)], \quad t \geq 0, \\ \hat{z}^{N_0}(t) &= 0, \quad t \leq 0. \end{aligned} \quad (47)$$

Recall $e^{N-N_0}(t)$ given in (19) and satisfying (24). Define $e^{N_0}(t) = z^{N_0}(t) - \hat{z}^{N_0}(t)$, where $z^{N_0}(t)$ is given in (12). The innovation term in (47) can be presented as

$$\begin{aligned} C_0 \hat{z}^{N_0}(t) + C_1 \hat{z}^{N-N_0}(t) - y(t) \\ = -C_0 e^{N_0}(t) - C_1 e^{N-N_0}(t) - \zeta(t) \end{aligned}$$

with $\zeta(t)$, given in (21), subject to (22). Using these notations with (6) and (47), we obtain

$$\dot{e}^{N_0}(t) = (A_0 - L_0C_0)e^{N_0}(t) - L_0C_1 e^{N-N_0}(t) - L_0\zeta(t). \quad (48)$$

As in [23], we propose the predictor-based control law

$$\bar{z}(t) = e^{A_0 r} \hat{z}^{N_0}(t) + \int_{t-r}^t e^{A_0(t-s)} B_0 u(s) ds, \quad u(t) = -K_0 \bar{z}(t) \quad (49)$$

Note that exponential decay of $\bar{z}(t)$ implies exponential decay of $\hat{z}^{N_0}(t)$ with the same decay rate. Differentiating $\bar{z}(t)$ and using (47) and (49) we obtain

$$\begin{aligned} \dot{\bar{z}}(t) &= (A_0 - B_0 K_0) \bar{z}(t) + e^{A_0 r} L_0 \\ &\times [C_0 e^{N_0}(t) + C_1 e^{N-N_0}(t) + \zeta(t)], \quad t \geq 0. \end{aligned} \quad (50)$$

Then, the reduced-order (decoupled from $\hat{z}^{N-N_0}(t)$) closed-loop system is given by non-delayed ODEs (24), (48), (50) and the tail

$$\dot{z}_n(t) = (-\lambda_n + q)z_n(t) - b_n K_0 \bar{z}(t-r), \quad n > N \quad (51)$$

which depends on the delay. Note also that in the case of state-feedback (see, e.g., [21]), the predictor is given by (49) with \hat{z}^{N_0} changed by z^{N_0} leading to decoupled from the tail ODE (50) with $L_0 = 0$. The latter simplifies the stability analysis of the closed-loop system and makes the proof of LMI feasibility trivial. Next, we consider L^2 -stability analysis of the closed-loop system, which is delay-independent for $\delta = 0$. Define the Lyapunov functional

$$\begin{aligned} \bar{V}(t) &= \bar{V}_0(t) + \left| e^{N_0}(t) \right|_{P_e}^2 + \int_{t-r}^t e^{-2\delta(t-s)} |\bar{z}(s)|_S^2 ds, \\ \bar{V}_0(t) &= |\bar{z}(t)|_{P_0}^2 + \sum_{n=N+1}^{\infty} z_n^2(t) + p_e \left| e^{N-N_0}(t) \right|^2, \end{aligned} \quad (52)$$

where $P_0, P_e, S > 0$ are matrices of appropriate dimensions and $p_e > 0$ is a scalar. By arguments similar to (31)-(38) the following LMIs guarantee $\dot{\bar{V}} + 2\delta \bar{V} \leq 0$ for $p_e \rightarrow \infty$:

$$\begin{aligned} -\epsilon_r S + \frac{2\alpha}{N\pi^2} K_0^T K_0 &< 0, \\ \begin{bmatrix} \psi + S & P_0 e^{A_0 r} L_0 C_0 & P_0 e^{A_0 r} L_0 \\ * & \psi_1 & -P_e L_0 \\ * & * & \psi_2 \end{bmatrix} &< 0, \\ \psi_1 &= P_e (A_0 - L_0 C_0) + (A_0 - L_0 C_0)^T P_e + 2\delta P_e, \\ \psi_2 &= \frac{2}{\|c\|_N^2} \left[-\lambda_{N+1} + q + \delta + \frac{1}{2\alpha} \lambda_{N+1} \right], \quad \epsilon_r = e^{-2\delta r}, \end{aligned} \quad (53)$$

TABLE I

SUB-PREDICTORS: MINIMAL N AND M FOR FEASIBILITY

r	0.2	0.4	0.6	0.8	1	1.1	1.2
N	4	6	12	22	41	62	90
M	1	4	7	10	12	13	18

TABLE II

CLASSICAL PREDICTOR: MINIMAL N FOR FEASIBILITY

r	0.5	1	1.5	2	2.3	2.5	2.8
N	7	12	19	34	42	58	88

where ψ is given in (36). Fix any $r > 0$. Let $\alpha = 1$, $S = \frac{1}{\sqrt{N}}I$, P_0 such that $\psi = -2I$ and P_e such that $\psi_1 = -\beta I$ for $\beta > 0$. Choosing first $\beta > 0$ large enough and then N large enough and applying Schur complement, (53) holds. Summarizing:

Proposition 2: Consider (3), measurement (4) with $c \in L^2(0, 1)$ satisfying (9) and control law (49). Let $\delta > 0$ be a desired decay rate. Let $N_0 \in \mathbb{Z}_+$ satisfy (7) and $N \geq N_0 + 1$. Let L_0 and K_0 be obtained using (10) and (11), respectively. Given $r > 0$, let there exist positive definite matrices P_0 , P_e , S and scalar $\alpha > 0$ such that LMIs (53) hold. Then the solution $z(x, t)$ to (3) under the control law (49) and the observer $\hat{z}(x, t)$ defined by (46) satisfy (39) for some constant $D > 0$. Furthermore, given any $r > 0$, the LMI (53) is feasible provided N is large enough.

C. Numerical Example

We consider (3) with $q = 3$, resulting in an unstable open-loop system. We consider (4) with $c(x) = \chi_{[0.3, 0.6]}$ (an indicator function). We fix $\delta = 0.1$ which results in $N_0 = 0$. The controller and observer gains are found using (10) and (11) as $K_0 = 8.8$, $L_0 = 14.66$.

We start with sub-predictors. Given various values of $r > 0$, the LMIs of Theorem 1 were verified for $1 \leq N+M \leq 110$ and $1 \leq M \leq 20$ by using the standard MATLAB LMI toolbox. Table I presents the minimal values of N and M found to guarantee the feasibility (i.e., the exponential stability of the closed-loop system with a decay rate 0.1).

For classical predictors, the LMIs of Proposition 2 were verified for $1 \leq N \leq 100$. Table II presents the minimal values of N which guarantee feasibility of the LMIs. It is seen from the tables that for the same values of r , the classical predictor employs a lower-order $N + 1$ -dimensional observer compared to $(M - 1)(N_0 + 1) + N + 1$ -dimensional sub-predictors. Numerical simulations appear in the arXiv version [28].

III. CONCLUSION

We studied constant input delay compensation by finite-dimensional observer-based controllers for the 1D heat equation. We proved that both sub-predictors and classical predictors theoretically compensate any delay provided the observer dimension is large. Classical predictors are known to be less friendly in application to uncertain systems (see, e.g., [19, Remark 3]). The suggested predictor methods can be extended in the future to various parabolic PDEs.

REFERENCES

[1] M. J. Balas, "Finite-dimensional controllers for linear distributed parameter systems: Exponential stability using residual mode filters," *J. Math. Anal. Appl.*, vol. 133, no. 2, pp. 283–296, 1988.

[2] P. Christofides, *Nonlinear and Robust Control of PDE Systems: Methods and Applications to Transport Reaction Processes*. New York, NY, USA: Birkhäuser, 2001.

[3] R. Curtain, "Finite-dimensional compensator design for parabolic distributed systems with point sensors and boundary input," *IEEE Trans. Autom. Control*, vol. AC-27, no. 1, pp. 98–104, Feb. 1982.

[4] C. Harkort and J. Deutscher, "Finite-dimensional observer-based control of linear distributed parameter systems using cascaded output observers," *Int. J. Control*, vol. 84, no. 1, pp. 107–122, 2011.

[5] T. Nambu, *Theory of Stabilization for Linear Boundary Control Systems*. Boca Raton, FL, USA: CRC Press, 2017.

[6] R. Katz and E. Fridman, "Constructive method for finite-dimensional observer-based control of 1-D parabolic PDEs," *Automatica*, vol. 122, Dec. 2020, Art. no. 109285.

[7] R. Katz and E. Fridman, "Finite-dimensional control of the Kuramoto-Sivashinsky equation under point measurement and actuation," in *Proc. 59th IEEE Conf. Decis. Control*, 2020, pp. 4423–4428.

[8] E. Fridman and A. Blighovsky, "Robust sampled-data control of a class of semilinear parabolic systems," *Automatica*, vol. 48, no. 5, pp. 826–836, 2012.

[9] N. Bar Am and E. Fridman, "Network-based H_∞ filtering of parabolic systems," *Automatica*, vol. 50, no. 12, pp. 3139–3146, 2014.

[10] I. Karafyllis and M. Krstic, "Sampled-data boundary feedback control of 1-D parabolic PDEs," *Automatica*, vol. 87, pp. 226–237, Jan. 2018.

[11] A. Selivanov and E. Fridman, "Delayed point control of a reaction-diffusion PDE under discrete-time point measurements," *Automatica*, vol. 96, pp. 224–233, Oct. 2018.

[12] R. Katz, E. Fridman, and A. Selivanov, "Boundary delayed observer-controller design for reaction-diffusion systems," *IEEE Trans. Autom. Control*, vol. 66, no. 1, pp. 275–282, Jan. 2021.

[13] R. Katz and E. Fridman, "Delayed finite-dimensional observer-based control of 1-D parabolic PDEs," *Automatica*, vol. 123, Jan. 2021, Art. no. 109364.

[14] Z. Artstein, "Linear systems with delayed controls: A reduction," *IEEE Trans. Autom. Control*, vol. 27, no. 4, pp. 869–879, Aug. 1982.

[15] M. Krstic, *Delay Compensation for Nonlinear, Adaptive, and PDE Systems*. Boston, MA, USA: Birkhäuser, 2009.

[16] A. Germani, C. Manes, and P. Pepe, "A new approach to state observation of nonlinear systems with delayed output," *IEEE Trans. Autom. Control*, vol. 47, no. 1, pp. 96–101, Jan. 2002.

[17] T. Ahmed-Ali, E. Cherrier, and F. Lamnabhi-Lagarrigue, "Cascade high gain predictors for a class of nonlinear systems," *IEEE Trans. Autom. Control*, vol. 57, no. 1, pp. 221–226, Jan. 2012.

[18] M. Najafi, S. Hosseinnia, F. Sheikholeslam, and M. Karimadini, "Closed-loop control of dead time systems via sequential sub-predictors," *Int. J. Control*, vol. 86, no. 4, pp. 599–609, 2013.

[19] Y. Zhu and E. Fridman, "Sub-predictors for network-based control under uncertain large delays," *Automatica*, vol. 123, Jan. 2021, Art. no. 109350.

[20] H. Sano, "Neumann boundary stabilization of one-dimensional linear parabolic systems with input delay," *IEEE Trans. Autom. Control*, vol. 63, no. 9, pp. 3105–3111, Sep. 2018.

[21] H. Lhachemi, C. Prieur, and R. Shorten, "An LMI condition for the robustness of constant-delay linear predictor feedback with respect to uncertain time-varying input delays," *Automatica*, vol. 109, Nov. 2019, Art. no. 108551.

[22] T. Ahmed-Ali, E. Fridman, F. Giri, M. Kahelras, F. Lamnabhi-Lagarrigue, and L. Burlion, "Observer design for a class of parabolic systems with large delays and sampled measurements," *IEEE Trans. Autom. Control*, vol. 65, no. 5, pp. 2200–2206, May 2020.

[23] R. Katz, E. Fridman, and I. Basre, "Delayed finite-dimensional observer-based control of 1D parabolic PDEs via reduced-order LMIs," *Automatica*, submitted for publication.

[24] S. Mondié and W. Michiels, "Finite spectrum assignment of unstable time-delay systems with a safe implementation," *IEEE Trans. Autom. Control*, vol. 48, no. 12, pp. 2207–2212, Dec. 2003.

[25] I. Karafyllis and M. Krstic, *Predictor Feedback for Delay Systems: Implementations and Approximations*. Basel, Switzerland: Birkhäuser, 2017.

[26] A. Pazy, *Semigroups of Linear Operators and Applications to Partial Differential Equations*, vol. 44. New York, NY, USA: Springer, 1983.

[27] E. Fridman, *Introduction to Time-Delay Systems: Analysis and Control (Systems and Control: Foundations and Applications)*. Cham, Switzerland: Birkhäuser, 2014.

[28] R. Katz and E. Fridman, "Sub-predictors and classical predictors for finite-dimensional observer-based control of parabolic PDEs," 2021. [Online]. Available: arXiv:3718703.